

# Industrial Control Part VII: Step-Response Analysis

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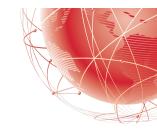




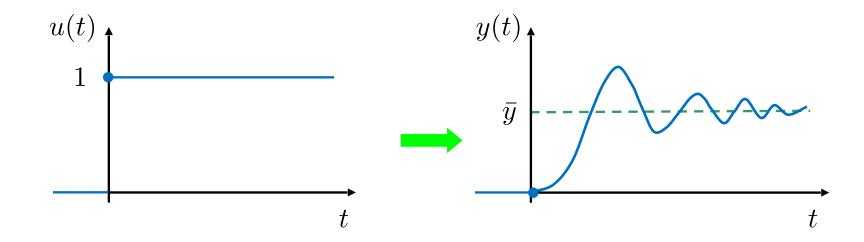




# **Step Response**



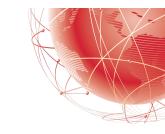
$$x(0) = 0; \quad u(t) = 1(t)$$



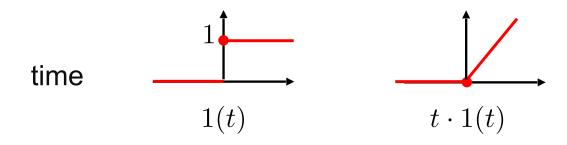


Recall that, for asymptotically stable systems, the step response describes the way the systems "moves" from an equilibrium to another

# Step Response (contd.)



The knowledge of the step response allows to easily determine the response to other inputs related to the step function by linear transformations



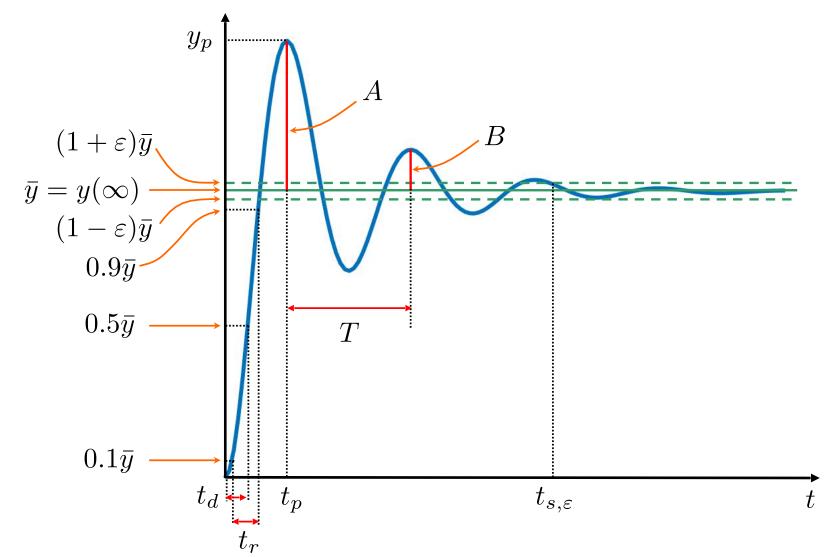
Laplace 
$$\frac{1}{s}$$

$$t \cdot 1(t) = \int_0^t 1(\tau)d\tau$$
 The response to the ramp function is the integral of the step-response

# **Characteristic Parameters of the Step Response**

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# Characteristic Parameters of the Step Response (contd.)



• Steady-state value:  $\bar{y} = y(\infty)$ 

• Settling time:  $t_{s,\varepsilon}$ 

• Rise time:  $t_r$ 

• Delay time:  $t_d$ 

• Peak time:  $t_p$ 

• Peak value:  $y_p$ 

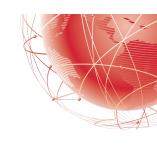
• Maximum overshoot:  $A = y_p - y(\infty)$ 

• Maximum percentage overshoot:  $\Delta\% = 100 \cdot A/y(\infty)$ 

"Period" of oscillations: T

• Damping factor: B/A

# **Step Response: First Order Systems**



# Case A)

$$G(s) = \frac{\mu}{1+s\tau}\,; \quad \mu > 0; \, \tau > 0 \qquad \text{strictly proper first-order system}$$
 asymptotic stability



# Case B)

$$G(s) = \frac{\mu(1+sT)}{1+s\tau}; \quad \mu > 0; \ \tau > 0 \quad \text{non strictly proper first-order system}$$



asymptotic stability

# Step Response: First Order Systems (contd.)

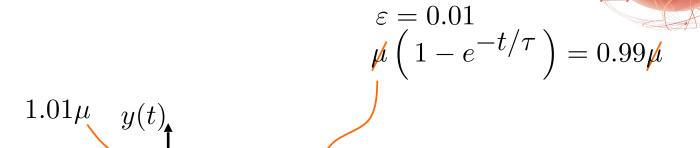


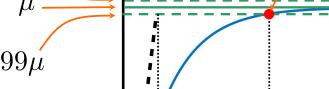
$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right]$$

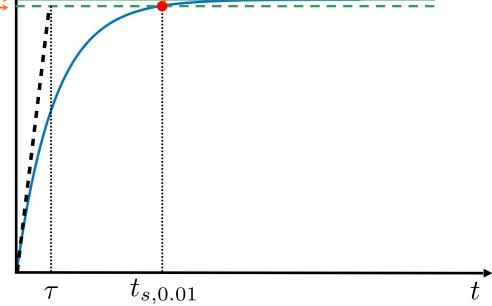
$$= \mathcal{L}^{-1} \left[ \frac{\mu}{s(1+s\tau)} \right] \qquad 0.99\mu'$$

$$= \mathcal{L}^{-1} \left[ \frac{\mu}{s} - \frac{\mu \tau}{1 + s \tau} \right]$$

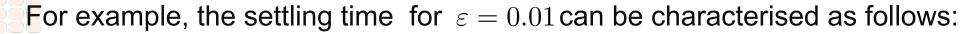
$$= \mu \left( 1 - e^{-t/\tau} \right) , \quad t \ge 0$$







# **Settling-Time Calculation**



$$1 - e^{-t/\tau} = 0.99 \implies e^{-t/\tau} = 0.01 \implies e^{t/\tau} = 100$$

$$t_{s,0.01} = \tau \ln 100 \simeq 4.6\tau$$

The calculation of the rising time  $\,t_r$  and the delay time  $\,t_d\,$  follows similar lines.  $\,t_{s,0.01}\,$ 

The following approximations are useful:

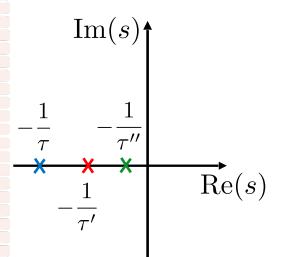
$$t_r \simeq 2.2\tau$$
  $t_d \simeq 0.7\tau$   $t_{s,0.05} \simeq 3\tau$   $t_{s,0.01} \simeq 4.6\tau$ 



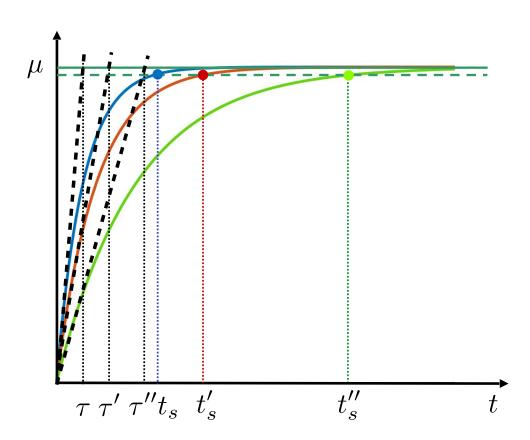
**Remark**: without loss of generality, from now on we shall use  $t_s$  as a shorthand for  $t_{s,0.01}$ 

# Qualitative Analysis of the Step Response









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# Step Response: First Order Systems (contd.)

# Case B)

$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right]$$

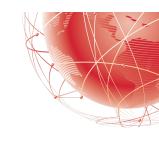
$$= \mathcal{L}^{-1} \left[ \frac{\mu(1+sT)}{s(1+s\tau)} \right]$$

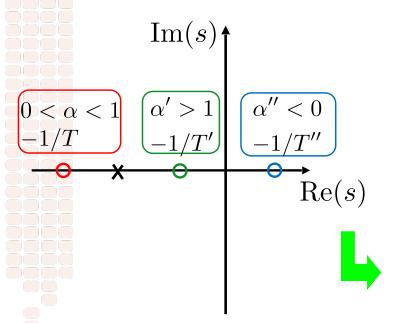
$$= \mathcal{L}^{-1} \left[ \frac{\mu}{s} + \frac{\mu(T-\tau)}{1+s\tau} \right]$$

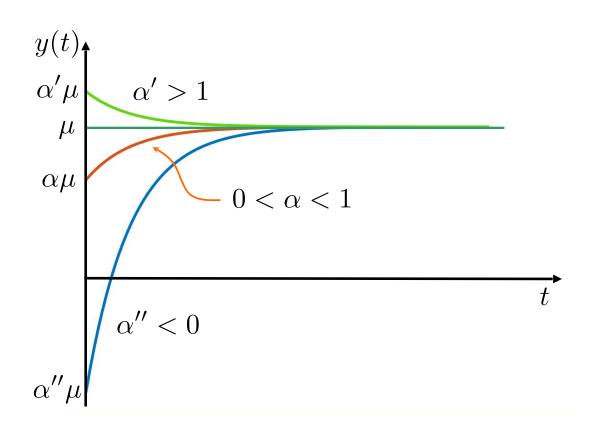
$$= \mu \left( 1 + (\alpha - 1)e^{-t/\tau} \right), \quad t \ge 0 \text{ with } T = \alpha \tau$$

Note that (the system is not strictly proper):  $\lim_{t\to 0^+}y(t)=\mu\frac{T}{\tau}\neq 0$ 

# Qualitative Analysis of the Step Response







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# **Step Response: Second Order Systems**



# Case A)

$$G(s) = \frac{\mu}{(1+s au_1)(1+s au_2)}$$
 real poles, no zeros

Case B)

$$G(s) = \frac{\mu(1+sT)}{(1+s\tau_1)(1+s\tau_2)}$$
 real poles, one zero

Case C)

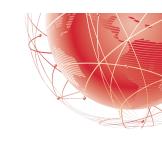
$$G(s) = \frac{\varrho}{(s + \sigma + j\omega)(s + \sigma - j\omega)}$$

complex poles, no zeros

Case D)

$$G(s) = \frac{\varrho(1+sT)}{(s+\sigma+j\omega)(s+\sigma-j\omega)}$$

complex poles, one zero



# • Case A)

$$G(s) = \frac{\mu}{(1+s\tau_1)(1+s\tau_2)}; \quad \mu > 0; \quad \tau_1 \neq \tau_2$$

$$\left. egin{array}{l} au_1 > 0 \ au_2 > 0 \end{array} 
ight. 
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m asymptotic stability$$

Without loss of generality, assume  $\tau_1 > \tau_2$ 



$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right] = \mathcal{L}^{-1} \left[ \frac{\mu}{s(1+s\tau_1)(1+s\tau_2)} \right]$$
$$= \mathcal{L}^{-1} \left[ \frac{A}{s} + \frac{B}{1+s\tau_1} + \frac{C}{1+s\tau_2} \right]$$

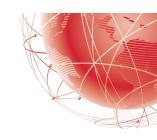
where

$$A = \frac{\mu}{(1+s\tau_1)(1+s\tau_2)}\Big|_{s=0} = \mu$$

$$B = \frac{\mu}{s(1+s\tau_2)}\Big|_{s=-1/\tau_1} = \frac{\mu}{-\frac{1}{\tau_1}(1-\frac{\tau_2}{\tau_1})} = \frac{\mu\tau_1^2}{\tau_2-\tau_1}$$

$$C = \frac{\mu}{s(1+s\tau_1)}\Big|_{s=-1/\tau_2} = \frac{\mu}{-\frac{1}{\tau_2}(1-\frac{\tau_1}{\tau_2})} = \frac{\mu\tau_2^2}{\tau_1-\tau_2}$$





#### Hence:

$$y(t) = \mathcal{L}^{-1} \left[ \frac{\mu}{s} - \frac{\frac{\mu \tau_1^2}{\tau_2 - \tau_1}}{1 + s\tau_1} + \frac{\frac{\mu \tau_2^2}{\tau_1 - \tau_2}}{1 + s\tau_2} \right]$$

$$= \mu \left( 1 - \frac{\tau_1}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \ge 0$$

#### **Characteristics:**

• 
$$y(0) = 0$$

$$\bullet \quad \dot{y}(0) = 0$$

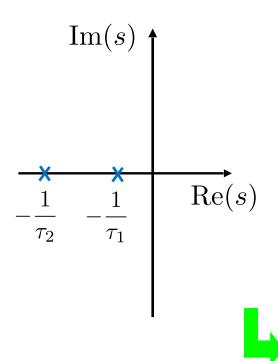
$$y(0) = 0$$

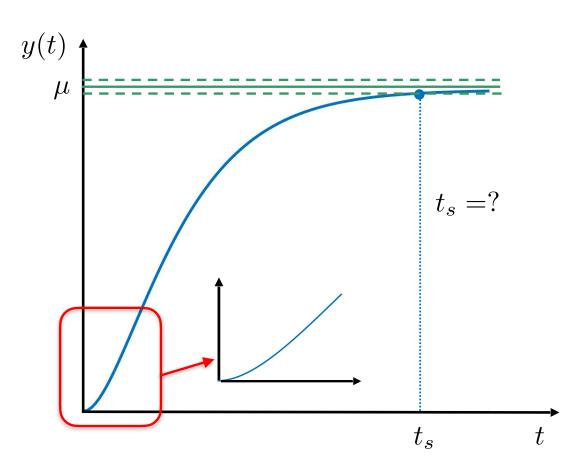
$$\dot{y}(0) = 0$$

$$\ddot{y}(0) = \frac{\mu}{\tau_1 \tau_2} > 0$$

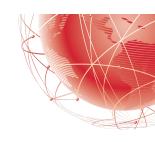
# Qualitative Analysis of the Step Response







# **Approximate Calculation of the Settling Time**



If  $\tau_1 \gg \tau_2$ :

$$y(t) = \mu \left( 1 - \frac{\tau_1}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \ge 0$$

$$\simeq \mu \left( 1 - e^{-t/\tau_1} \right), \quad t \ge 0$$

$$t_s \simeq 4.6\tau_1$$

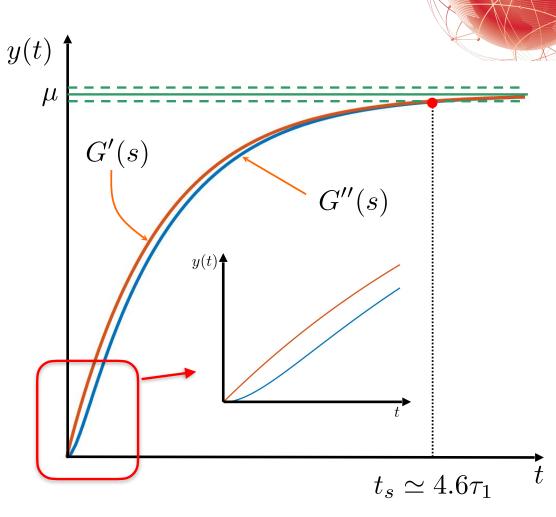
In general, in the absence of zeros, the most influential poles on the qualitative behaviour of the step response are the ones closer to the imaginary axis.

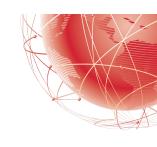
# Qualitative Analysis: Comparison Between First and Second Order

$$G'(s) = \frac{\mu}{1 + s\tau_1}$$

$$G''(s) = \frac{\mu}{(1 + s\tau_1)(1 + s\tau_2)}; \ \tau_1 \gg \tau_2$$

- The main difference lies in the initial transient behaviour
- For a given settling time, the stepresponse in the second-order case without zeros has a "slower" dynamics

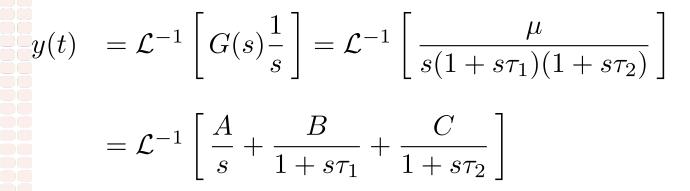




# • Case B)

$$G(s) = \frac{\mu(1+sT)}{(1+s\tau_1)(1+s\tau_2)}; \quad \mu > 0; \quad \tau_1 \neq \tau_2$$

Without loss of generality, assume  $\tau_1 > \tau_2$ 

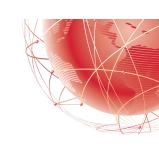


where

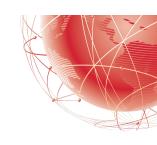
$$A = \frac{\mu(1+sT)}{(1+s\tau_1)(1+s\tau_2)} \Big|_{s=0} = \mu$$

$$B = \frac{\mu(1+sT)}{s(1+s\tau_2)} \Big|_{s=-1/\tau_1} = \frac{\mu(1-T/\tau_1)}{-\frac{1}{\tau_1}(1-\frac{\tau_2}{\tau_1})} = \frac{\mu\tau_1(\tau_1-T)}{\tau_2-\tau_1}$$

$$C = \frac{\mu(1+sT)}{s(1+s\tau_1)} \Big|_{s=-1/\tau_2} = \frac{\mu(1-T/\tau_2)}{-\frac{1}{\tau_2}(1-\frac{\tau_1}{\tau_2})} = \frac{\mu\tau_2(\tau_2-T)}{\tau_1-\tau_2}$$



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#### Hence:

$$y(t) = \mu \left( 1 - \frac{\tau_1 - T}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{\tau_2 - T}{\tau_1 - \tau_2} e^{-t/\tau_2} \right), \quad t \ge 0$$

#### Characteristics:

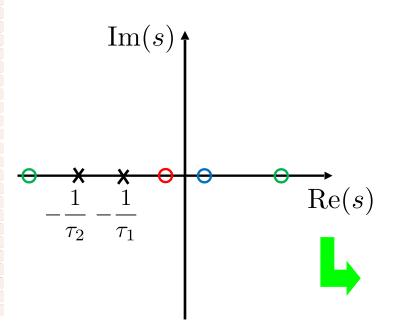
$$\bullet \quad y(\infty) = \mu > 0$$

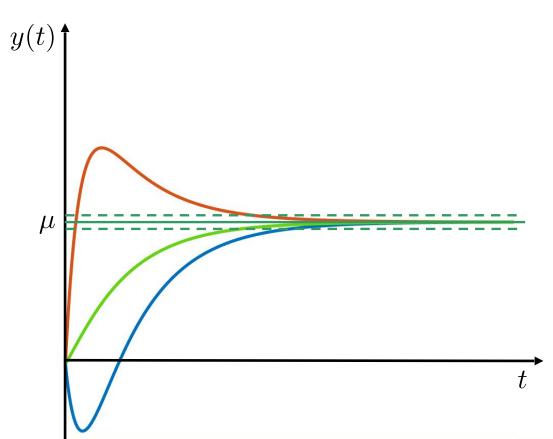
$$\bullet \quad y(0) = 0$$

• 
$$y(0) = 0$$
  
•  $\dot{y}(0) = \frac{\mu T}{\tau_1 \tau_2}$   $\begin{cases} > 0, & \text{if } T > 0 \\ < 0, & \text{if } T < 0 \end{cases}$ 

# **Qualitative Analysis of the Step Response**







- zero with little influence
- overshoot
- undershoot



# Case C)

$$G(s) = \frac{\varrho}{(s + \sigma + j\omega)(s + \sigma - j\omega)}$$

$$\mu = G(0) = \frac{\varrho}{\sigma^2 + \omega^2}$$

poles: 
$$-\sigma \pm j\omega$$

$$\sigma > 0$$
 asymptotic stability

$$\omega > 0$$

$$\varrho > 0$$



$$Y(s) = \frac{G(s)}{s} = \frac{A}{s} + \frac{Bs + C}{s^2 + 2\sigma s + \sigma^2 + \omega^2}$$



$$As^2 + 2A\sigma s + A\sigma^2 + A\omega^2 + Bs^2 + Cs = \varrho$$



$$\begin{cases} A + B = 0 \\ 2A\sigma + C = 0 \\ A(\sigma^2 + \omega^2) = \varrho \end{cases} \longrightarrow \begin{cases} A = \frac{\varrho}{\sigma^2 + \omega^2} = \mu \\ B = -\mu \\ C = -2\sigma\mu \end{cases}$$

$$\begin{cases} A = \frac{\varrho}{\sigma^2 + \omega^2} = \mu \\ B = -\mu \\ C = 2\sigma\mu \end{cases}$$



$$Y(s) = \mu \left[ \frac{1}{s} - \frac{s + 2\sigma}{s^2 + 2\sigma s + \sigma^2 + \omega^2} \right] = \mu \left[ \frac{1}{s} - \frac{s + \sigma + \sigma}{(s + \sigma)^2 + \omega^2} \right]$$

$$= \mu \left[ \frac{1}{s} - \frac{s+\sigma}{(s+\sigma)^2 + \omega^2} - \frac{\sigma}{\omega} \frac{\omega}{(s+\sigma)^2 + \omega^2} \right]$$



Hence: 
$$y(t) = \mu \left[ 1 - e^{-\sigma t} \cos(\omega t) - \frac{\sigma}{\omega} e^{-\sigma t} \sin(\omega t) \right], \quad t \ge 0$$

$$= \mu \left[ 1 - e^{-\sigma t} \left( \cos(\omega t) + \frac{\sigma}{\omega} \sin(\omega t) \right) \right], \quad t \ge 0$$

$$= \mu \left[ 1 - \frac{\sqrt{\sigma^2 + \omega^2}}{\omega} e^{-\sigma t} \sin(\omega t + \varphi) \right], \quad t \ge 0$$

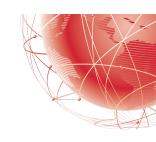
damped oscillations

where 
$$\varphi = \arccos\left(\frac{\sigma}{\sqrt{\sigma^2 + \omega^2}}\right)$$



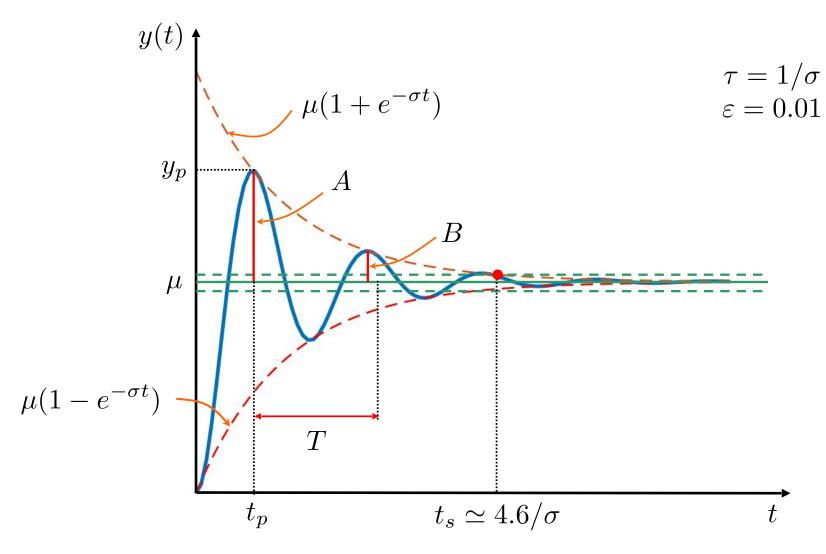
- Characteristics:  $\bullet$   $y(\infty) = \mu > 0$ 
  - $\bullet \quad y(0) = 0$
  - $\bullet \quad \dot{y}(0) = 0$

• 
$$\ddot{y}(0) = \varrho > 0$$



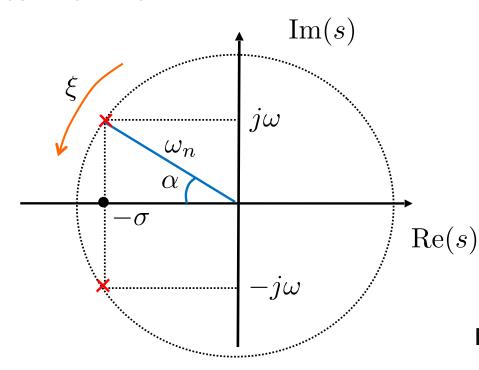
# **Qualitative Analysis of the Step Response**





# **Characteristic Parameters of the Step Response**

#### Recall from Part 4:



$$G(s) = \frac{\varrho}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

$$\omega_n^2 = \sigma^2 + \omega^2$$

$$\omega_n \xi = \sigma$$

$$\omega_n \sqrt{1 - \xi^2} = \omega$$

#### Parameters:

 $\omega_n$  natural angular frequency:

 $\xi = \cos(\alpha)$  damping ratio



# Characteristic Parameters of the Step Response (contd.)

and:

$$G(s) = \frac{\varrho}{(s+\sigma+j\omega)(s+\sigma-j\omega)} = \frac{\varrho}{(s+\sigma)^2 + \omega^2}$$
$$= \frac{\varrho}{s^2 + 2\sigma s + \sigma^2 + \omega^2} = \frac{\varrho}{s^2 + 2\xi\omega_n s + \omega_n^2}$$
$$2\xi\omega_n \qquad \omega_n^2$$

$$G(s) = \frac{\varrho/\omega_n^2}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2} = \frac{\mu}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}$$

where: 
$$\mu := \frac{\varrho}{\omega_n^2}$$

# Characteristic Parameters of the Step Response (contd.)

### Hence:

• Settling time: 
$$t_s \simeq \frac{5}{\sigma} = \frac{4.6}{\xi \omega_n}$$

• Peak time: 
$$t_p = \frac{\pi}{\omega} = \frac{\pi}{\omega_n \sqrt{1-\xi^2}}$$

• Peak value: 
$$y_p = \mu \left[1 + e^{-\frac{\sigma\pi}{\omega}}\right] = \mu \left[1 + e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}}\right]$$

• Maximum percentage overshoot: 
$$\Delta\% = 100 \cdot \frac{A}{\mu} = e^{-\sigma\pi/\omega} = 100 \cdot e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}}$$

• "Period" of oscillations: 
$$T=\frac{2\pi}{\omega}=\frac{2\pi}{\omega_n\sqrt{1-\xi^2}}$$
• Damping factor: 
$$\frac{B}{A}=\cdots=\Delta^2=e^{-2\sigma\pi/\omega}=e^{-\frac{2\xi\pi}{\sqrt{1-\xi^2}}}$$

• Damping factor: 
$$\frac{B}{A}=\cdots=\Delta^2=e^{-2\sigma\pi/\omega}=e^{-\frac{2\xi\pi}{\sqrt{1-\xi^2}}}$$

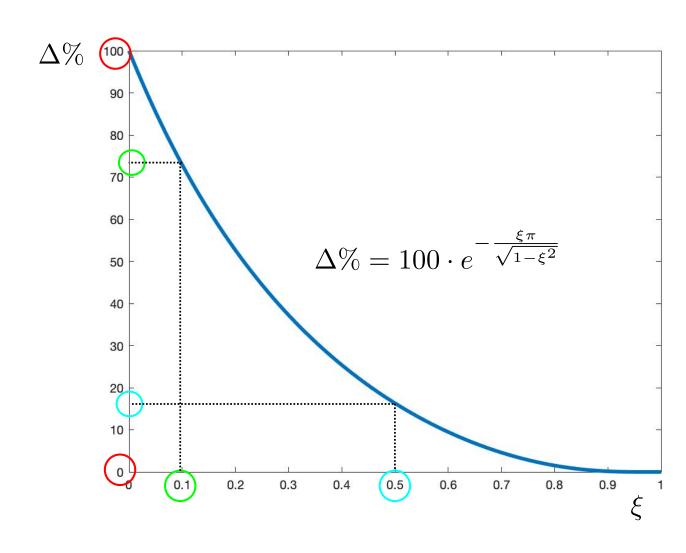
only depend on  $\xi$ but **not** on  $\omega_n$ 



# **Maximum Percentage Overshoot**



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# **Limit Cases**



• No damping:  $\xi = 0$ 

$$G(s) = \frac{\varrho}{s^2 + \omega_n^2}$$
 poles:  $\pm j\omega_n$ 

poles: 
$$\pm j\omega_n$$



Undamped oscillations

Full damping:  $\xi = 1$ 

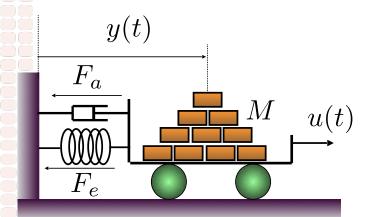
$$G(s) = \frac{\varrho}{(s + \omega_n)^2}$$
 poles:  $-\omega_n$ ;  $-\omega_n$ 

poles: 
$$-\omega_n; \, -\omega_n$$



No oscillations at all

# **Example 1**



Hence:

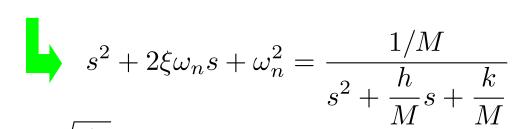
$$\mu = G(0) = \frac{1}{k}$$

$$2\xi\omega_n = \frac{h}{M}$$

$$\omega_n^2 = \frac{k}{M}$$

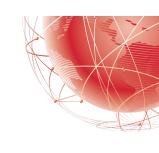


$$G(s) = \frac{1}{Ms^2 + hs + k}$$
$$= \frac{1/M}{s^2 + \frac{h}{M}s + \frac{k}{M}}$$



$$\omega_n = \sqrt{\frac{k}{M}}$$

$$\xi = \frac{h}{2\sqrt{kM}}$$



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# **Example 2**



$$\begin{array}{ccc}
 & \underline{x_2} & \underline{x_2} & \underline{x_2} & \underline{x_1} & \underline{x_2} &$$

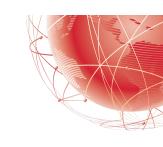
$$\begin{cases}
C\dot{x}_1 = u - x_2 \\
L\dot{x}_2 = x_1 - Rx_2 \\
y = Rx_2
\end{cases}$$

$$A = \begin{bmatrix} 0 & -1/C \\ 1/L & -R/L \end{bmatrix}$$

$$A = \begin{bmatrix} 0 & -1/C \\ 1/L & -R/L \end{bmatrix} \qquad B = \begin{bmatrix} 1/C \\ 0 \end{bmatrix} \qquad C = \begin{bmatrix} 0 & R \end{bmatrix}$$

$$G(s) = \begin{bmatrix} 0 & R \end{bmatrix} \begin{bmatrix} s & 1/C \\ -1/L & s+R/L \end{bmatrix}^{-1} \begin{bmatrix} 1/C \\ 0 \end{bmatrix} = \dots = \frac{R/(LC)}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

$$\omega_n = \frac{1}{\sqrt{LC}}; \quad \xi = \frac{R}{2} \sqrt{\frac{C}{L}}; \quad \mu = R$$



# Case D)

$$G(s) = \frac{\mu(1+sT)}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}; \quad 0 < \xi < 1; \ \omega_n > 0; \ \mu > 0$$

Characteristics of the step response:

• 
$$y(\infty) = \mu > 0$$

• 
$$y(0) = 0$$

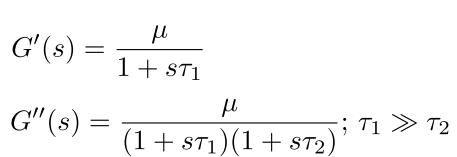
$$y(\infty) = \mu > 0$$

$$y(0) = 0$$

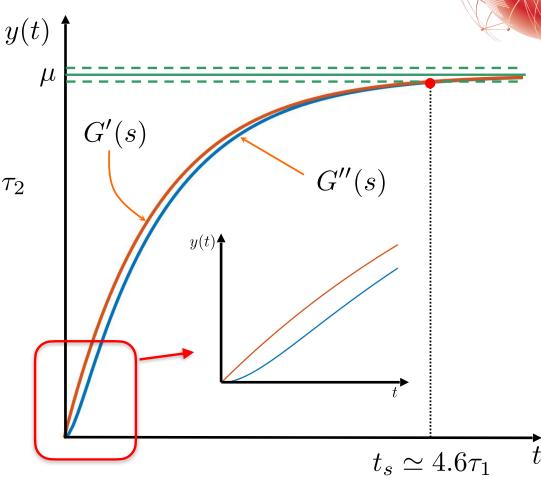
$$\dot{y}(0) = \mu T \omega_n^2$$

$$< 0, \quad \text{if } T < 0$$

# Qualitative Analysis: Comparison between first and second order



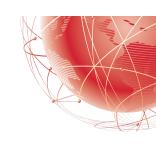
- The main difference lies in the init transient behaviour
- For a given settling time, the step response in the second-order cas without zeros has a "slower" dynamics



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# **Qualitative Analysis:**

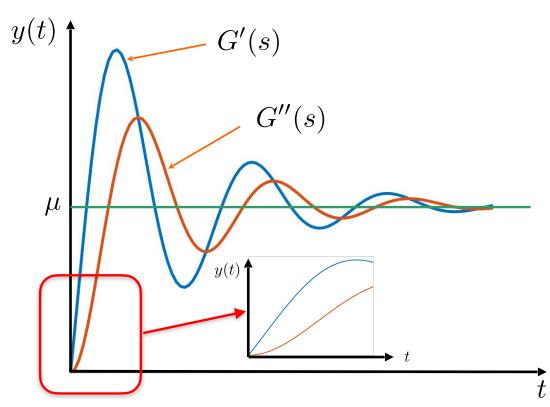
# Comparison between Case C) (no zeros) and Case D) (one zero)



$$G'(s) = \frac{\mu}{1 + \frac{2\xi}{\omega_n} s + \frac{1}{\omega_n^2} s^2} \qquad y(t) \uparrow$$

$$G''(s) = \frac{\mu(1+sT)}{1 + \frac{2\xi}{\omega_n}s + \frac{1}{\omega_n^2}s^2}$$

- Again, the main difference lies in the initial transient behaviour
- For a given settling time, the stepresponse in Case C) without zeros has a "slower" dynamics



# **Step Response for Systems of Order > 2**



For simplicity, consider the case of real poles only:

$$G(s) = \frac{\mu}{s^g} \frac{\prod_{i=1}^{m} (1 + sT_i)}{\prod_{i=1}^{n} (1 + s\tau_i)}$$

Recall (in the absence of common factors in G(s)):

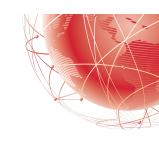
**Asymptotic Stability** 



$$Re(poles) < 0$$
$$g < 0$$

$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right]$$

# **Step Response for Systems of Order > 2 (contd.)**

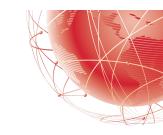


#### Initial Value Theorem

## Final Value Theorem

$$\lim_{t \to \infty} y(t) = \lim_{s \to 0} \operatorname{sf} \frac{1}{s} G(s) \begin{cases} = \mu, & \text{if } g = 0 \\ = 0, & \text{if } g < 0 \end{cases}$$

# **Dominant Poles Approximation**



Again, for simplicity, consider the case of real poles:

$$Y(s) = G(s)\frac{1}{s} = \frac{\alpha_0}{s} + \frac{\alpha_1}{1 + s\tau_1} + \dots + \frac{\alpha_n}{1 + s\tau_n}$$

$$y(t) = \mathcal{L}^{-1} \left[ G(s) \frac{1}{s} \right]$$

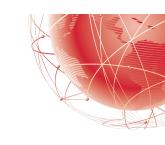
$$= \alpha_0 + \frac{\alpha_1}{\tau_1} e^{-t/\tau_1} + \dots + \frac{\alpha_n}{\tau_n} e^{-t/\tau_n}$$

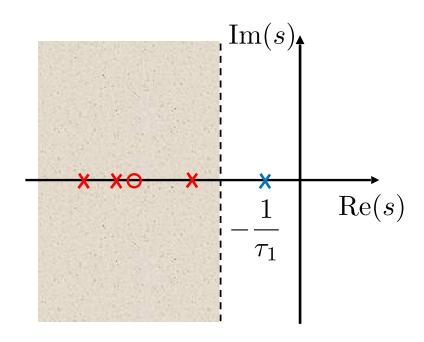
Assuming:  $\tau_1 > \tau_2 > \cdots > \tau_n$ 

$$y(t) = \alpha_0 + \frac{\alpha_1}{\tau_1} e^{-t/\tau_1} + \dots + \frac{\alpha_n}{\tau_n} e^{-t/\tau_n}$$

$$\simeq \alpha_0 + rac{lpha_1}{ au_1} e^{-t/ au_1}$$
 dominant component, hence:  $t_s \simeq 5 au_1$ 

# **Dominant Poles Approximation: Real Poles**

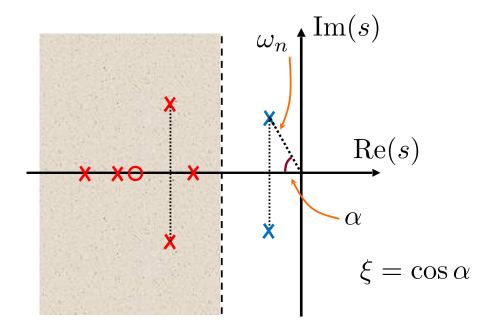






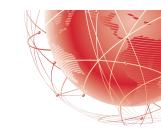
# **Dominant Poles Approximation: Real and Complex Poles**







# **Example**



$$G(s) = \frac{400(1+s)}{(1+0.2s)(1+0.1s)(s^2+2s+4)}$$

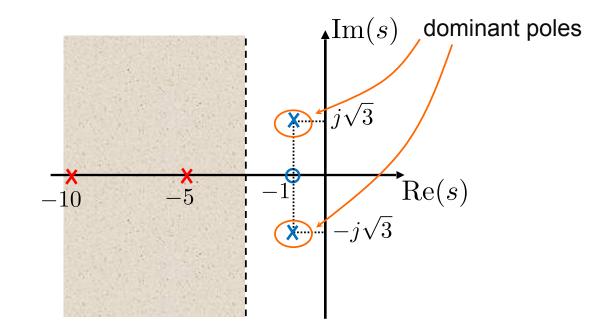
$$\psi_n = 2$$

$$\xi = 1/2$$

$$\mu = G(0) = 100$$

poles: 
$$-5$$
  
 $-10$   
 $-1 \pm j\sqrt{3}$ 

zero: -1

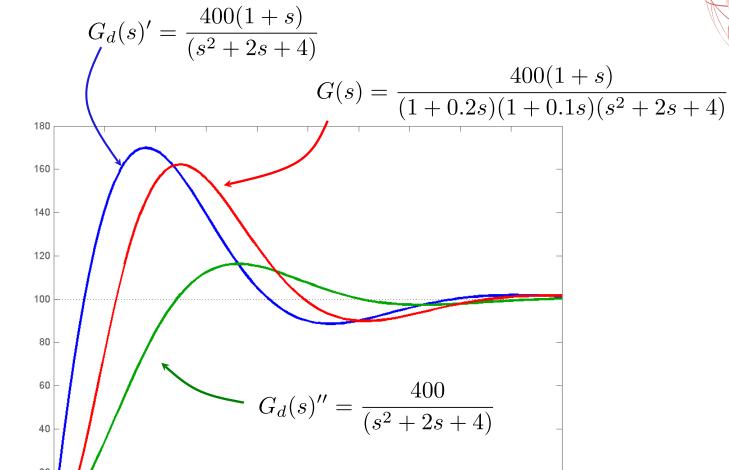


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# Example (contd.)





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2

2.5

3.5

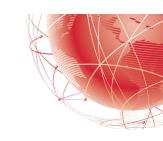
4.5

43

1.5

0.5

# **Equivalent Pole Approximation**



Suppose that all poles are real and negative:

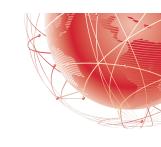
$$G(s) = \frac{\mu}{(1+s\tau_1)(1+s\tau_2)\cdots(1+s\tau_n)}; \quad \tau_i > 0, \ i = 1,\dots, n$$

$$G_e(s) = \frac{\mu}{1 + s\tau_e}, \qquad \tau_e := \sum_{i=1}^n \tau_i$$



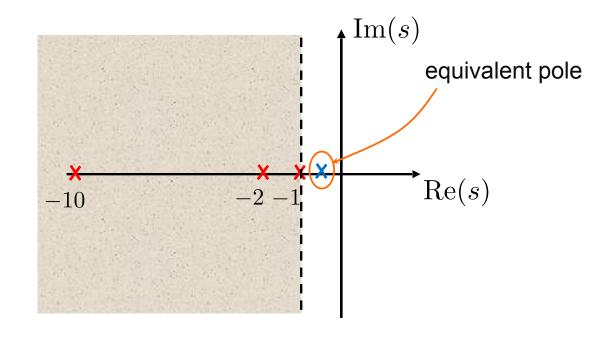
dominant component, hence:  $t_s \simeq 5\tau_e$ 

# **Example**



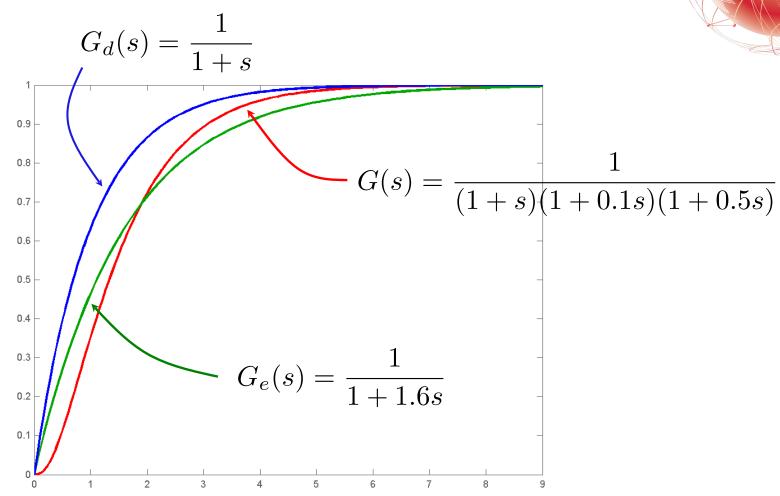
$$G(s) = \frac{1}{(1+s)(1+0.1s)(1+0.5s)}$$

$$\mu = G(0) = 1$$
 poles:  $-1$   $-2$   $-10$ 



# Example (contd.)

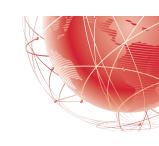






# Kolos

# **Dominant and Equivalent Poles Approximation: Remarks**



- When using the dominant poles approximation:
  - It is important to "preserve" the gain
  - Zeros located close to the imaginary axis have to be properly taken into account
- The equivalent pole approximation can only be used when all poles are real and negative
- Both approximations are useful in qualitative analysis and the for initial controller's design steps